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# S-72.333 Postgraduate course in radio communications

Channel Capacity, MIMO capacity

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## Outline

- Introduction
- Definitions of Channel and Channel Capacity
- Capacity of Discrete Memoryless Channel
- Capacity of Gaussian Channel
- MIMO capacity
- Summary and conclusions

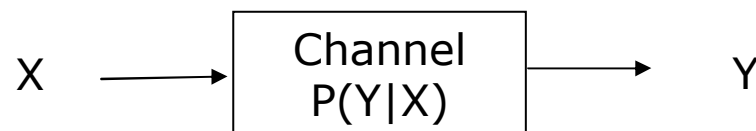


## Introduction

- Channel capacity is a fundamental limit on the rate of error free messages that can take place over a channel
- Advantage of Multiple-Input Multiple-Output (MIMO) systems is increased capacity



## Definitions (Shannon)



- **Channel is defined by conditional distribution of channel output given input**
  - Channel input and output are random variables where distribution of the input determines distribution of the output
  - Mutual information,  $I(X;Y)$ , can be considered as amount of the information conveyed through channel. It depends on the input distribution.
- **Channel capacity is maximum mutual information of input and output**

$$C = \max_{p(x)} I(X; Y)$$

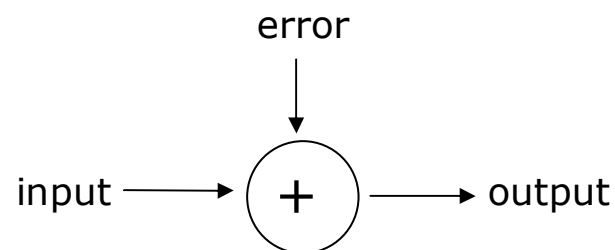


## Example: Binary Symmetric Channel

In Binary Symmetric Channel (BSC)  
 **$output = input + error \text{ mod } 2$**   
where  **$input$** ,  **$output$**  and  **$error$**  are  
*binary random variables taking  
values 0 or 1.*

Bit error probability is

$$P_e = P(error = 1)$$

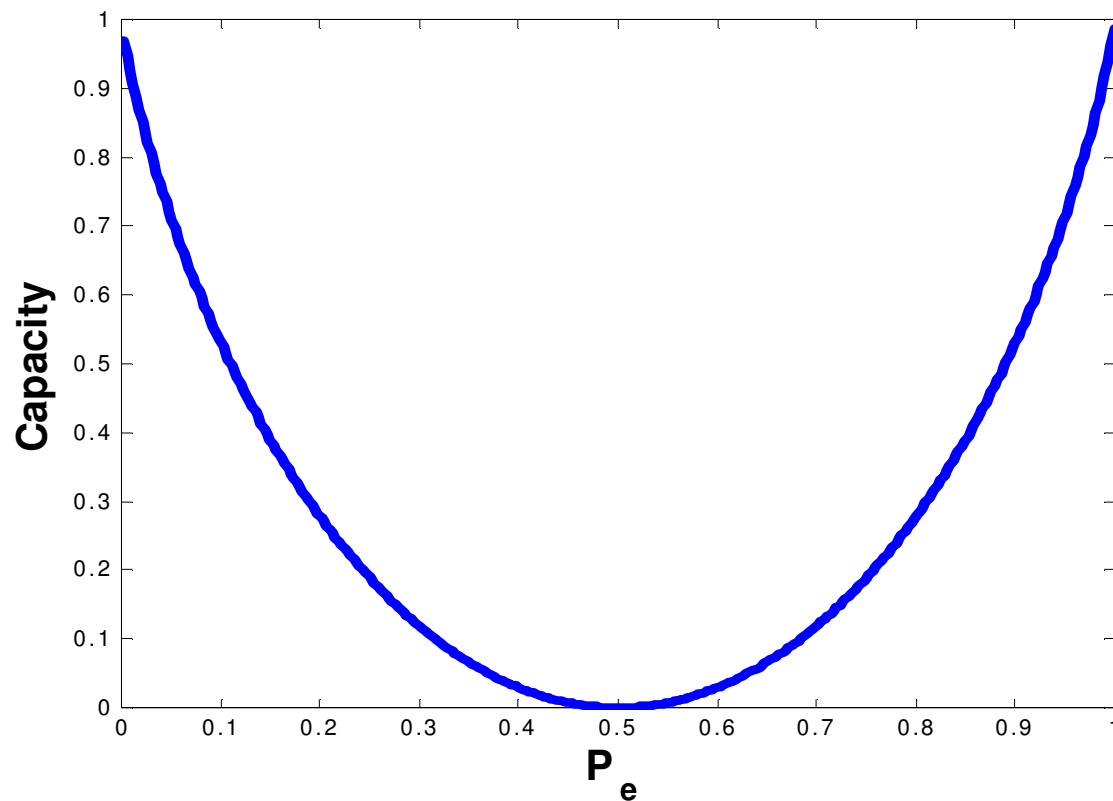


$$P(output | input) = \begin{bmatrix} 1 - p_e & p_e \\ p_e & 1 - p_e \end{bmatrix}$$



## Example: Binary Symmetric Channel

$$C_{BSC} = 1 + p_e \log_2 p_e + (1 - p_e) \log_2 (1 - p_e)$$

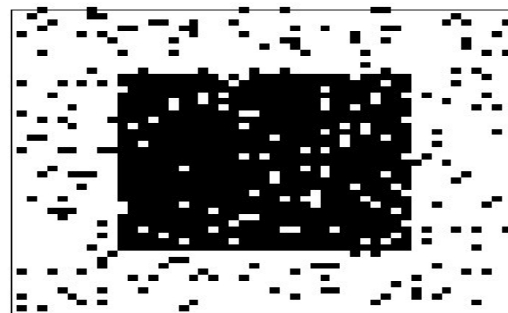




## Example: Binary Symmetric Channel



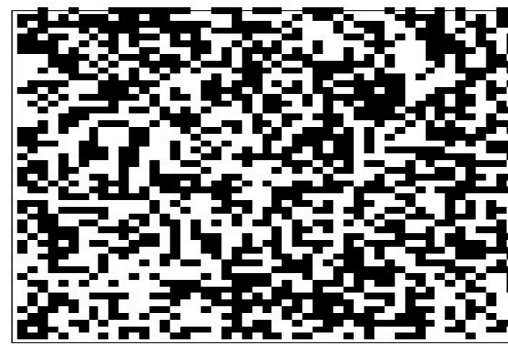
$P_e = 0, C = 1$



$P_e = 0.1, C = 0.5$



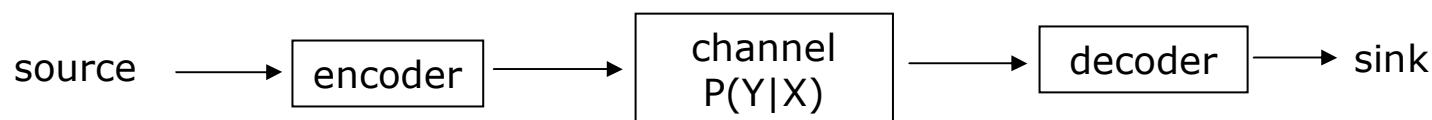
$P_e = 0.9, C = 0.5$



$P_e = 0.5, C = 0$



### Channel Coding Theorem



- Channel coding theorem (Shannon 1948) says that reliable information rates up to channel capacity are possible.
- Channel coding theorem do not say how to achieve these reliable rates. Efforts on designing efficient channel codes has been going on ever since.
- Theorem do not promise non-erroneous transmissions when application has delay constraints. Result of the theorem is asymptotic: non-erroneus transmissions require unlimited block lengths.





### Capacity of Gaussian Channel

- Gaussian channel is time discrete continuous alphabet channel. Output is sum of the input and a Gaussian noise, where input and noise are independent.

$$Y = X + Z \quad Z \sim \mathbf{N}(0, N)$$

- Since input and noise are independent, mutual information of output and input is differential entropy of output minus differential entropy of noise.

$$\begin{aligned} I(X; Y) &= h(Y) - h(Y | X) \\ &= h(Y) - h(Z | X) \\ &= h(Y) - h(Z) \end{aligned}$$



### Capacity of Gaussian Channel

- Without restrictions on input, capacity would be infinity.
- With constraint  $E(X^2) \leq P$  mutual information is maximized when input is Gaussian with variance  $P$ . Then output is a sum of two independent Gaussians and so it too is Gaussian, with variance  $N+P$ .
- Differential entropy of Gaussian random variable with variance  $\sigma^2$  is  $\frac{1}{2} \log_2(2\pi e \sigma^2)$ , thus

$$\begin{aligned} C_G &= \frac{1}{2} \log_2(2\pi e(N+P)) - \frac{1}{2} \log_2(2\pi eN) \\ &= \frac{1}{2} \log_2\left(1 + \frac{P}{N}\right) \end{aligned}$$



### Capacity of bandlimited Gaussian Channel

- Bandlimited Gaussian channel is a continuous time channel

$$Y = (X + Z) * h_w$$

where where Y is output, X is input, Z is white noise and  $h_w$  is impulse response of perfect bandpass filter.

- Output can be reconstructed from samples taken at  $1/2W$  interval. Capacity of a sample is achieved when input is Gaussian

$$C_{sample} = \frac{1}{2} \log_2 \left( 1 + \frac{P}{WN_0} \right) \text{ bits/sample}$$

- $2W$  samples are taken per second, so channel capacity per second is  $2W$  times sample capacity

$$C = 2WC_{sample} = W \log_2 \left( 1 + \frac{P}{2WN_0} \right) \text{ bits/second}$$



### Capacity of RF channel

- In general we can approximate capacity of the RF channel with equation

$$C = \log_2(1 + \rho h^2) \text{ bps/Hz}$$

where  $\rho$  is *SNR* and  $h$  is channel impulse response.



## MIMO channel

- General formula for MIMO channel is

$$Y = H S + n$$

- $H$  is  $N \times M$  matrix where  $M$  ( $N$ ) is the number of input (output) antennas. Elements  $h_{ij}$  of  $H$  represent channel impulse responses between  $i$ th input and  $j$ th output antenna
- $S$  is  $M \times 1$  random variable representing input antennas
- $Y$  is  $N \times 1$  random variable representing output antennas

$$H(\tau, t) = \begin{bmatrix} h_{1,1}(\tau, t) & h_{1,2}(\tau, t) & \cdots & h_{1,M}(\tau, t) \\ h_{2,1}(\tau, t) & h_{2,2}(\tau, t) & \cdots & h_{2,M}(\tau, t) \\ \vdots & \vdots & \ddots & \vdots \\ h_{N,1}(\tau, t) & h_{N,2}(\tau, t) & \cdots & h_{N,M}(\tau, t) \end{bmatrix}$$



## Capacity of deterministic MIMO channel

- In deterministic MIMO channel matrix is assumed to be non-random
- $n$  is additive temporally white complex zero-mean Gaussian noise with covariance matrix

$$E(nn^H) = N_0 I_M$$

- $S$  is power constrained  $\sum_{i=1}^M E(s_i^2) \leq P$
- Mutual information of output and input is

$$\begin{aligned} I(Y; S) &= h(Y) - h(Y | S) \\ &= h(Y) - h(n | S) \\ &= h(Y) - h(n) \end{aligned}$$

- Covariance matrix of  $Y$  is

$$\begin{aligned} E(YY^H) &= E(HS)E(n)^H + E(HS(HS)^H) + E(n)E(HS)^H + E(nn^H) \\ &= E(HSS^H H^H) + N_0 I_M \\ &= HR_{SS}H^H + N_0 I_M \end{aligned}$$

where  $R$  is the covariance matrix of  $S$



### Capacity of deterministic MIMO channel

- Mutual information of input and output is maximized when input is Gaussian => output is Gaussian.
- Differential entropy of multivariate Gaussian is  $\frac{1}{2} \log_2((2\pi e)^n |K|)$  where  $|K|$  is determinant of the covariance matrix

$$C^D = \log_2(|I_M + \frac{\rho}{M} HR_{SS}H^H|) \text{ bps/Hz}$$

- In the absence of channel state information at the transmitter, it is reasonable to choose  $S$  to be spatially white i.e.  $R = I$ . Then we have

$$C^D = \log_2(|I + \frac{\rho}{M} HH^H|)$$
$$= \sum_{i=1}^r \log_2(1 + \frac{\rho}{M} \lambda_i) \text{ bps/Hz}$$

where  $r$  is the rank of and  $\lambda_i$  denote positive eigenvalues of  $HH^H$ .



### SISO, MISO, SIMO and MIMO capacities

- For a deterministic channel matrix without channel knowledge at the transmitter we have

$$C_{SISO}^D = \log_2(1 + \rho h^2) \text{ bps/Hz}$$

$$C_{MISO}^D = \log_2\left(1 + \frac{\rho}{M} \sum_{i=1}^M h_i^2\right) \text{ bps/Hz}$$

$$C_{SIMO}^D = \log_2\left(1 + \rho \sum_{i=1}^N h_i^2\right) \text{ bps/Hz}$$

$$C_{MIMO}^D = \sum_{i=1}^r \log_2\left(1 + \frac{\rho}{M} \lambda_i\right) \text{ bps/Hz}$$

- SIMO capacity increases more rapidly with number of receiving antennas than MISO capacity with number of transmitting antennas. This is because MISO has no spatial array gain at the receiver.





### MIMO capacity

- At it's best MIMO capacity increase linearly with  $\min(M,N)$
- Advantageous of MIMO depends on  $H$ ; the larger rank and eigenvalues  $HH^H$  have, the more MIMO capacity we have.



### Ergodic Capacity of Gaussian MIMO channel with Rayleigh fading

- Additive noise is Gaussian and entries of channel matrix are complex gaussian i.i.d zero mean unit variance random variables.
- Channel matrix is a random variable => mutual information of input and output is a random variable=> "channel capacity" is a random variable **C**.
- Definition: **ergodic capacity**  $eC$  is a mean of **C**,  $eC = E(\mathbf{C})$ .
- **ergodic capacity** is a function of  $E(\mathbf{H}\mathbf{H}^H)$
- By choosing  $S$  to be i.i.d Gaussian and noting that entries of  $\mathbf{H}\mathbf{H}^H$  are  $\chi_{2M}^2$  at the diagonal and zero-mean else, we get

$$eC = C^D$$

where  $C^D$  is capacity of the deterministic MIMO channel whose channel matrix has  $\mathbf{H}\mathbf{H}^H = E(\chi_{2M}^2) \mathbf{I}_N$



### Outage Probability

- Channel coding theorem do not apply with ergodic capacity i.e. ergodic capacity do not express maximal zero-error information rate.
- Definition: **Outage probability** for a given rate  $R$  is probability that mutual information falls below that rate,

$$P_{out}(R) = P(C \leq R)$$

- Outage probability can be interpreted as PER.



### Capacity of frequency-selective fading MIMO channels

- MIMO capacity of frequency flat channel is high
- Frequency band can be divided into subchannels so that bandwidth of subchannels is less than coherence bandwidth => subchannel is flat
- Assuming that transmit power is allocated uniformly across transmit antennas, mutual information is given by

$$I_{FS} = \frac{1}{K} \sum_{i=1}^K \log_2 \left| I_M + \frac{\rho}{N} H_i H_i^H \right| \text{ bps/Hz}$$

and ergodic capacity is  $E(I_{FS})$

- Outage probability will in general be lower when frequency band is divided



### Summary and conclusions

- Main ideas about deriving channel capacities for discrete, Gaussian, RF and MIMO channels were presented.
- Capacity of RF channel can be approximated by logarithmic function of SNR
- MIMO capacity depends on channel matrix. At it's best MIMO capacity may increase linearly with number of antennas.



### References

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## Homeworks

- (1) Is it possible to increase channel capacity by channel coding?
- (2) Find channel capacities for deterministic multiple antenna channels when SNR = 10dB (=10) and channel matrices are

(a)  $H = [1]$  (SISO)

(b)  $H = \begin{bmatrix} 1 \\ 1 \\ 1 \end{bmatrix}$  (SIMO)

(c)  $H = [1 \ 1 \ 1]$  (MISO)

(d)  $H = \begin{bmatrix} 1 & 0 & 0 \\ 0 & 1 & 0 \\ 0 & 0 & 1 \end{bmatrix}$  (MIMO) (rank(H) = 3, all three eigenvalues are 1)