

S-72.341 CODING METHODS

Tutorial 7, Solutions

1. The rows of the generator matrix are given in Example 7-3 (Figure 7-1). The parity-check matrix is the generator matrix of the dual code of $\mathfrak{R}(2,4)$, i.e. $\mathfrak{R}(1,4)$. This is a subcode of $\mathfrak{R}(2,4)$ (see problem 4.2). The generator matrix of $\mathfrak{R}(1,4)$ consists of the first five rows of Figure 7-1, that is

$$\mathbf{H}_{\mathfrak{R}(2,4)} = \mathbf{G}_{\mathfrak{R}(1,4)} = \begin{bmatrix} 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 1 & 1 & 1 & 1 \\ 0 & 0 & 0 & 0 & 1 & 1 & 1 & 1 & 0 & 0 & 0 & 0 & 1 & 1 & 1 \\ 0 & 0 & 1 & 1 & 0 & 0 & 1 & 1 & 0 & 0 & 1 & 1 & 0 & 0 & 1 \\ 0 & 1 & 0 & 1 & 0 & 1 & 0 & 1 & 0 & 1 & 0 & 1 & 0 & 1 & 0 \end{bmatrix}.$$

2. This problem is about basic properties of RM codes.

a) Dimension $k = 1 + \binom{5}{1} + \binom{5}{2} = 16$. See equation (7-2).

b) Rate of the code is $\frac{16}{32} = \frac{1}{2}$.

c) $d_{\min} = 2^3 = 8$. $t = \left\lfloor \frac{8-1}{2} \right\rfloor = 3$. See Theorem 7-2.

- d) This part has nothing to do with RM codes, but it reminds us of the basic principles of decoding of block codes (chapter 4 in the book).

Since probability of a bit error is 0.5, all 32-tuples are equally probable at the receiver. Thus, we need to calculate the probability of the received vector being within a Hamming sphere of a wrong code word. There are $2^{16} - 1$ wrong code words. Hamming spheres of radius 3 have volume

$$V_2(32, 3) = \sum_{j=0}^3 \binom{32}{j} = 5489.$$

Probability of *decoding error* is

$$P_{de} = \frac{(2^{16} - 1) \cdot 5489}{2^{32}} \approx 0.084.$$

The probability of *decoding failure* is

$$P_{df} = \frac{2^{32} - 2^{16} \cdot 5489}{2^{32}} \approx 0.916.$$

Note: For a bounded-distance decoder there is a difference between decoder error and decoder failure, see page 74 of the book.

The probability that a code word is decoded correctly is only $5489/2^{32} \approx 1.3 \cdot 10^{-6}$. In this case “decoding” is perhaps not a very meaningful term, since all 32-tuples are equally probably.

3. A $\mathfrak{R}(m-1, m)$ code has a dual code $\mathfrak{R}(0, m)$. Generator matrix of the $\mathfrak{R}(0, m)$ code (a row vector of ones) is the parity-check matrix \mathbf{H} of the $\mathfrak{R}(m-1, m)$ code. The condition

$$\mathbf{c}^T \mathbf{H} = \mathbf{c}^T \underbrace{\begin{bmatrix} 1 & 1 & \cdots & 1 \end{bmatrix}^T}_{\text{even \# of ones}} = 0$$

must be satisfied all code words \mathbf{c} . This means that all 2^{2^m-1} code words must have even weight. Thus, the code constitutes all even-weight 2^m -tuples.

4. We need to compute the checksums for message bit estimates m_1, m_2, m_3 , and m_4 . The algorithm for doing this is described on pp. 160-161 of the course book. Here computation for obtaining the checksums for m_1 is shown. Others follow similarly.

- i. The incidence vector is $\mathbf{v}_1 = 01010101010101$, see Example 7-3. The subspace associated with \mathbf{v}_1 is

$$\begin{aligned} S &= \{P_1, P_3, P_5, P_7, P_9, P_{11}, P_{13}, P_{15}\} \\ &= \{(1110), (1100), (1010), (1000), (0110), (0100), (0010), (0000)\}. \end{aligned}$$

Table 7-2 on page 161 can be used to obtain the 4-tuples.

- ii. The complementary subspace of S can be found by examining the incidence vector $\mathbf{v}_2 \mathbf{v}_3 \mathbf{v}_4 = 0000000000000011$. This results in

$$T = \{P_{14}, P_{15}\} = \{(0001), (0000)\}.$$

- iii. The first checksum is $m_1^1 = c_{14} + c_{15}$.

- iv. The translation of T wrt $P_1 = (1110)$ is $\{(1111), (1110)\} = \{P_0, P_1\}$. Notice that all translations result in the original point in the subspace S and a point whose index is lowered by 1. Thus, the remaining translations are $\{P_3, P_2\}$, $\{P_5, P_4\}$, $\{P_7, P_6\}$, $\{P_9, P_8\}$, $\{P_{11}, P_{10}\}$, $\{P_{13}, P_{12}\}$, and $\{P_{15}, P_{14}\}$. For message bit m_1 the orthogonal checksums m_1^i , $i = 1, 2, \dots, 8$, are

$$\begin{aligned} m_1^1 &= c_{14} + c_{15} \\ m_1^2 &= c_{12} + c_{13} \\ m_1^3 &= c_{10} + c_{11} \\ m_1^4 &= c_8 + c_9 \\ m_1^5 &= c_6 + c_7 \\ m_1^6 &= c_4 + c_5 \\ m_1^7 &= c_2 + c_3 \\ m_1^8 &= c_0 + c_1 \end{aligned}$$

Similarly for the other message bits we obtain

$$\begin{array}{lll} m_2^1 = c_{13} + c_{15} & m_3^1 = c_{11} + c_{15} & m_4^1 = c_7 + c_{15} \\ m_2^2 = c_{12} + c_{14} & m_3^2 = c_{10} + c_{14} & m_4^2 = c_6 + c_{14} \\ m_2^3 = c_9 + c_{11} & m_3^3 = c_9 + c_{13} & m_4^3 = c_5 + c_{13} \\ m_2^4 = c_8 + c_{10} & m_3^4 = c_8 + c_{12} & m_4^4 = c_4 + c_{12} \\ m_2^5 = c_5 + c_7 & m_3^5 = c_3 + c_7 & m_4^5 = c_3 + c_{11} \\ m_2^6 = c_4 + c_6 & m_3^6 = c_2 + c_6 & m_4^6 = c_2 + c_{10} \\ m_2^7 = c_1 + c_3 & m_3^7 = c_1 + c_5 & m_4^7 = c_1 + c_9 \\ m_2^8 = c_0 + c_2 & m_3^8 = c_0 + c_4 & m_4^8 = c_0 + c_8. \end{array}$$

5. The Reed decoding algorithm is explained in the course book, starting from page 155. In the following we decode the first 16-tuple, and give only answer to the remaining three.

Before the actual decoding we must find the checksums for each message bit. Checksums for 2nd order message bits m_{12} , m_{13} , m_{34} are given in equations (7-8) and (7-9) of the book. To obtain checksums for m_{14} , m_{24} , and m_{34} we examine the tesseracts in Figures 7-3a and 7-3b.

Checksums for m_{14} : Using the lower left tesseract in Figure 7-3a we identify the following complementary subspaces:

$$\begin{aligned} T_{14}^1 &= \{(0000), (0001), (1001), (1000)\} = \{P_0, P_1, P_9, P_8\} \\ T_{14}^2 &= \{(0010), (1010), (1011), (0011)\} = \{P_2, P_{10}, P_{11}, P_3\} \\ T_{14}^3 &= \{(0100), (0101), (1101), (1100)\} = \{P_4, P_5, P_{13}, P_{12}\} \\ T_{14}^4 &= \{(0110), (1110), (0111), (1111)\} = \{P_6, P_{14}, P_7, P_{15}\}. \end{aligned}$$

The checksums are:

$$\begin{aligned} m_{14}^1 &= c_0 + c_1 + c_8 + c_9 \\ m_{14}^2 &= c_2 + c_3 + c_{10} + c_{11} \\ m_{14}^3 &= c_4 + c_5 + c_{12} + c_{13} \\ m_{14}^4 &= c_6 + c_7 + c_{14} + c_{15}. \end{aligned}$$

Checksums for m_{24} : Using the middle tesseract in Figure 7-3a we identify the following complementary subspaces:

$$\begin{aligned} T_{24}^1 &= \{(0000), (1000), (1010), (0010)\} = \{P_0, P_8, P_{10}, P_2\} \\ T_{24}^2 &= \{(0001), (1001), (1011), (0011)\} = \{P_1, P_9, P_{11}, P_3\} \\ T_{24}^3 &= \{(0101), (0111), (1111), (1101)\} = \{P_5, P_7, P_{15}, P_{13}\} \\ T_{24}^4 &= \{(0110), (1110), (1100), (0100)\} = \{P_6, P_{14}, P_{12}, P_4\}. \end{aligned}$$

The checksums are:

$$\begin{aligned} m_{24}^1 &= c_0 + c_2 + c_8 + c_{10} \\ m_{24}^2 &= c_1 + c_3 + c_9 + c_{11} \\ m_{24}^3 &= c_5 + c_7 + c_{13} + c_{15} \\ m_{24}^4 &= c_4 + c_6 + c_{12} + c_{14}. \end{aligned}$$

Checksums for m_{23} : Using the uppermost tesseract in Figure 7-3b we identify the following complementary subspaces:

$$\begin{aligned} T_{23}^1 &= \{(0000), (0010), (0100), (0110)\} = \{P_0, P_2, P_4, P_6\} \\ T_{23}^2 &= \{(0001), (0011), (0111), (0101)\} = \{P_1, P_3, P_7, P_5\} \\ T_{23}^3 &= \{(1001), (1101), (1011), (1111)\} = \{P_9, P_{13}, P_{11}, P_{15}\} \\ T_{23}^4 &= \{(1000), (1010), (1100), (1110)\} = \{P_8, P_{10}, P_{12}, P_{14}\}. \end{aligned}$$

The checksums are:

$$\begin{aligned} m_{23}^1 &= c_0 + c_2 + c_4 + c_6 \\ m_{23}^2 &= c_1 + c_3 + c_5 + c_7 \\ m_{23}^3 &= c_9 + c_{11} + c_{13} + c_{15} \\ m_{23}^4 &= c_8 + c_{10} + c_{12} + c_{14}. \end{aligned}$$

The checksums for the 1st order terms were calculated in problem 4.

- a) First we compute estimates for second order message bits by substitution $c_i \rightarrow r_i$, $i = 0, 1, 2, \dots, 15$ in the checksums. The received word is

$$\mathbf{r} = \begin{array}{cccccccccccccccc} 0 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 1 \\ r_0 & r_1 & r_2 & r_3 & r_4 & r_5 & r_6 & r_7 & r_8 & r_9 & r_{10} & r_{11} & r_{12} & r_{13} & r_{14} & r_{15} \end{array}$$

Here we go:

$$\begin{aligned} \hat{m}_{12}^1 &= r_0 + r_1 + r_2 + r_3 = 0 + 0 + 1 + 1 = 0 \\ \hat{m}_{12}^2 &= r_4 + r_5 + r_6 + r_7 = 0 + 1 + 1 + 1 = 1 \\ \hat{m}_{12}^3 &= r_8 + r_9 + r_{10} + r_{11} = 0 + 0 + 0 + 1 = 1 \\ \hat{m}_{12}^4 &= r_{12} + r_{13} + r_{14} + r_{15} = 0 + 1 + 1 + 1 = 1 \end{aligned}$$

$$\begin{aligned} \hat{m}_{13}^1 &= r_0 + r_1 + r_4 + r_5 = 1 \\ \hat{m}_{13}^2 &= r_2 + r_3 + r_6 + r_7 = 0 \\ \hat{m}_{13}^3 &= r_8 + r_9 + r_{12} + r_{13} = 1 \\ \hat{m}_{13}^4 &= r_{10} + r_{11} + r_{14} + r_{15} = 1 \end{aligned}$$

$$\begin{aligned} \hat{m}_{14}^1 &= r_0 + r_1 + r_8 + r_9 = 0 \\ \hat{m}_{14}^2 &= r_2 + r_3 + r_{10} + r_{11} = 1 \\ \hat{m}_{14}^3 &= r_4 + r_5 + r_{12} + r_{13} = 0 \\ \hat{m}_{14}^4 &= r_6 + r_7 + r_{14} + r_{15} = 0 \end{aligned}$$

$$\begin{aligned} \hat{m}_{23}^1 &= r_0 + r_2 + r_4 + r_6 = 0 \\ \hat{m}_{23}^2 &= r_1 + r_3 + r_5 + r_7 = 1 \\ \hat{m}_{23}^3 &= r_9 + r_{11} + r_{13} + r_{15} = 1 \\ \hat{m}_{23}^4 &= r_8 + r_{10} + r_{12} + r_{14} = 1 \end{aligned}$$

$$\hat{m}_{24}^1 = r_0 + r_2 + r_8 + r_{10} = 1$$

$$\hat{m}_{24}^2 = r_1 + r_3 + r_9 + r_{11} = 0$$

$$\hat{m}_{24}^3 = r_5 + r_7 + r_{13} + r_{15} = 0$$

$$\hat{m}_{24}^4 = r_4 + r_6 + r_{12} + r_{14} = 0$$

$$\hat{m}_{34}^1 = r_0 + r_4 + r_8 + r_{12} = 0$$

$$\hat{m}_{34}^2 = r_1 + r_5 + r_9 + r_{13} = 0$$

$$\hat{m}_{34}^3 = r_2 + r_6 + r_{10} + r_{14} = 1$$

$$\hat{m}_{34}^4 = r_3 + r_7 + r_{11} + r_{15} = 0$$

Next we estimate the message bits based on majority vote:

$$\hat{m}_{12} = \text{maj}\{0,1,1,1\} = 1$$

$$\hat{m}_{13} = \text{maj}\{1,0,1,1\} = 1$$

$$\hat{m}_{14} = \text{maj}\{0,1,0,0\} = 0$$

$$\hat{m}_{23} = \text{maj}\{0,1,1,1\} = 1$$

$$\hat{m}_{24} = \text{maj}\{1,0,0,0\} = 0$$

$$\hat{m}_{34} = \text{maj}\{0,0,1,0\} = 0.$$

Estimated second order message bits of the message word $\mathbf{m} = [\mathbf{m}_0 \mid \mathbf{m}_1 \mid \mathbf{m}_2]$ are

$$\begin{aligned} \hat{\mathbf{m}}_2 &= [\hat{m}_{34} \quad \hat{m}_{24} \quad \hat{m}_{14} \quad \hat{m}_{23} \quad \hat{m}_{13} \quad \hat{m}_{12}] \\ &= [0 \quad 0 \quad 0 \quad 1 \quad 1 \quad 1]. \end{aligned}$$

The effect of the 2nd order terms is removed from the received word:

$$\mathbf{r}' = \mathbf{r} - \hat{\mathbf{m}}_2 \mathbf{G}_2.$$

The generator matrix is given in Figure 7-1. We get

$$\mathbf{r}' = \begin{bmatrix} 0 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 1 \end{bmatrix} -$$

$$\begin{bmatrix} 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 1 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 1 & 1 \\ 0 & 0 & 0 & 0 & 0 & 1 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 1 & 0 & 1 \\ 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 & 0 & 0 & 0 & 1 \end{bmatrix}$$

$$= - \frac{\begin{array}{cccccccccccccccc} 0 & 0 & 1 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 1 \\ 0 & 0 & 0 & 1 & 0 & 1 & 1 & 1 & 0 & 0 & 0 & 1 & 0 & 1 & 1 & 1. \end{array}}{\begin{array}{cccccccccccccccc} 0 & 0 & 1 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 & 0 \end{array}}$$

First order terms are obtained as

$$\hat{m}_1^1 = r'_{14} + r'_{15} = 0$$

$$\hat{m}_1^2 = r'_{12} + r'_{13} = 0$$

$$\hat{m}_1^3 = r'_{10} + r'_{11} = 0$$

$$\hat{m}_1^4 = r'_8 + r'_9 = 0$$

$$\hat{m}_1^5 = r'_6 + r'_7 = 0$$

$$\hat{m}_1^6 = r'_4 + r'_5 = 0$$

$$\hat{m}_1^7 = r'_2 + r'_3 = 1$$

$$\hat{m}_1^8 = r'_0 + r'_1 = 0$$

$$\hat{m}_2^1 = r'_{13} + r'_{15} = 0$$

$$\hat{m}_2^2 = r'_{12} + r'_{14} = 0$$

$$\hat{m}_2^3 = r'_9 + r'_{11} = 0$$

$$\hat{m}_2^4 = r'_8 + r'_{10} = 0$$

$$\hat{m}_2^5 = r'_5 + r'_7 = 0$$

$$\hat{m}_2^6 = r'_4 + r'_6 = 0$$

$$\hat{m}_2^7 = r'_1 + r'_3 = 0$$

$$\hat{m}_2^8 = r'_0 + r'_2 = 1$$

... and so on. Since \mathbf{r}' has weight one and all first order estimates are sums of two distinct received coordinates, all first order message bit estimates become zero under majority-vote decoding i.e. $\hat{m}_i = 0$, $i = 1, 2, 3, 4$. Thus,

$$\mathbf{r}'' = \mathbf{r}' - \underbrace{\hat{\mathbf{m}}_1 \mathbf{G}_1}_{=0} = \mathbf{r}'.$$

The estimate for the message bit m_0 is

$$\hat{m}_0 = \text{maj}\{r_0'', r_1'', \dots, r_{15}''\} = 0.$$

The decoded message is

$$\begin{aligned} \hat{\mathbf{m}} &= [\hat{m}_0 \quad \hat{m}_4 \quad \hat{m}_3 \quad \hat{m}_2 \quad \hat{m}_1 \quad \hat{m}_{34} \quad \hat{m}_{24} \quad \hat{m}_{14} \quad \hat{m}_{23} \quad \hat{m}_{13} \quad \hat{m}_{12}] \\ &= [0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0 \quad 1 \quad 1 \quad 1]. \end{aligned}$$

b) $\hat{\mathbf{m}} = [1 \quad 1 \quad 0 \quad 1 \quad 0 \quad 1 \quad 0 \quad 0 \quad 0 \quad 0 \quad 0].$

c) Decoder failure: $\hat{\mathbf{m}}_{34} = \text{maj}\{1, 0, 0, 1\}.$

d) $\hat{\mathbf{m}} = [0 \quad 1 \quad 0 \quad 0 \quad 1 \quad 1 \quad 0 \quad 0 \quad 1 \quad 0 \quad 0].$