

CDMA as Signal Space Waveform I

- SF chips (T_c) combined into symbol (duration $T = SF \cdot T_c$)
- signal-space waveform $s(t) = \sum_{n=1}^{SF} c_n f(t - nT_c)$
 - $f(t)$ is the pulse shape, most of power concentrated inside chip duration T_c
 - $\mathbf{c} = [c_1 \ c_2 \ \dots \ c_{SF}]^T$ is the spreading code
 - most power concentrated inside symbol duration $[0, T]$
 - $s(t)$ modulated with different complex symbol each period T
- if same entity transmits multiple symbols in $[0, T]$, *multicode* transmission used
 - multiple orthogonal waveforms in $[0, T]$
 - orthogonal spreading codes

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Multicode CDMA

- signal space model of transmitted chips:

$$\mathbf{s} = \begin{bmatrix} c_{11} & c_{12} & \dots & c_{1M} \\ c_{21} & c_{22} & \dots & c_{2M} \\ \vdots & \vdots & \ddots & \vdots \\ c_{SF,1} & c_{SF,2} & \dots & c_{SF,M} \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_M \end{bmatrix} \equiv \mathbf{C}\mathbf{x}$$

- M symbols x_m transmitted simultaneously, using M spreading codes
- each column in the $SF \times M$ spreading matrix \mathbf{C} is a spreading code
- with orthogonal codes, $\mathbf{C}^H \mathbf{C} = \mathbf{I}_M$

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CDMA as Signal Space Waveform II

- Bandwidth B occupied by CDMA waveform is $\sim 1/T_c$
 - for sinc pulses, $B = 1/T_c$ exactly
 - for raised cosine pulses with roll-off factor α , $B = (1 + \alpha)/T_c$
- The frequency domain corresponding to a finite time (or periodic) signal $s(t)$ is discrete, consists of Fourier series coefficients at frequencies k/T

$$f_k = \int_0^T dt s(t) e^{2\pi j kt/T}$$

- frequencies arising from transitions at symbol borders ignored
- CDMA signal $s(t)$ has non-vanishing frequencies $|k| \geq SF/2$
- full power density spectrum: F-transform of $s(t)\text{rect}(0, T)$

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CDMA as Signal Space Waveform III

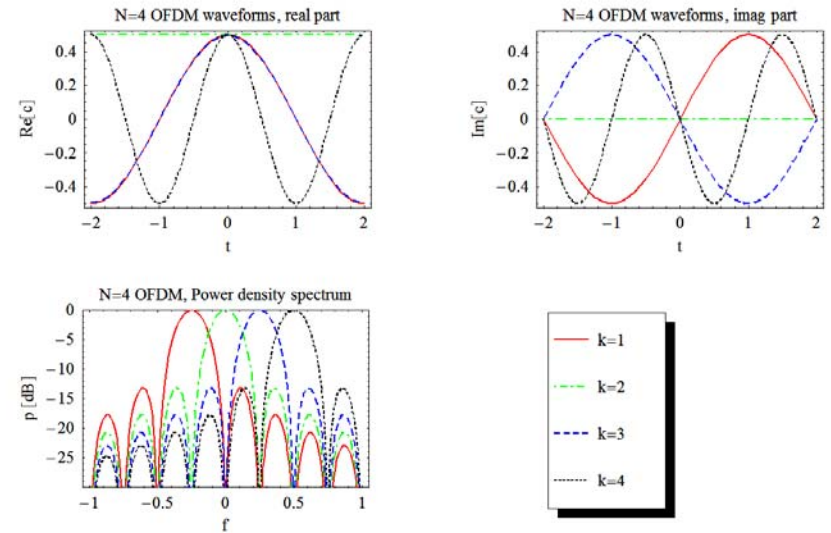
- the SF orthogonal spreading codes constitute a basis in the space of complex periodic functions with period T and bandwidth B
 - any function within $[0, T]$ and Fourier series components vanishing outside $[-B/2, B/2]$ can be expressed with orthogonal spreading codes and complex coefficients (the symbols)
 - elements c_i in Walsh-Hadamard spreading codes are real, but generic spreading code may have complex elements
- in time interval $[0, T]$ and bandwidth $B = 1/T_c$, there are $TB = T/T_c = SF$ orthogonal resources

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Fourier Basis

- Another natural basis: $\{f_k(t) = e^{2\pi j(N/2-k)t/NT_s}\}_{k=1}^N$
 - Periodic in $[0, T]$ (or $[-T/2, T/2]$), symbol duration $T = NT_s$
 - the Fourier sampling time T_s analogous to T_c
 - Fourier transform size N analogous to SF
 - Fourier bandwidth $B = 1/T_s$
 - $f_k(t)$ is *subcarrier*, bandwidth $B_s = 1/NT_s$
 - * constant amplitude, rotating phase—complex valued waveform
 - * localized in frequency $[k - 1/2, k + 1/2]/NT_s - 1/2T_s$
- any function within $[0, T]$ and Fourier series components vanishing outside $[-B/2, B/2]$ can be expressed with subcarriers waveforms and complex coefficients (the symbols)

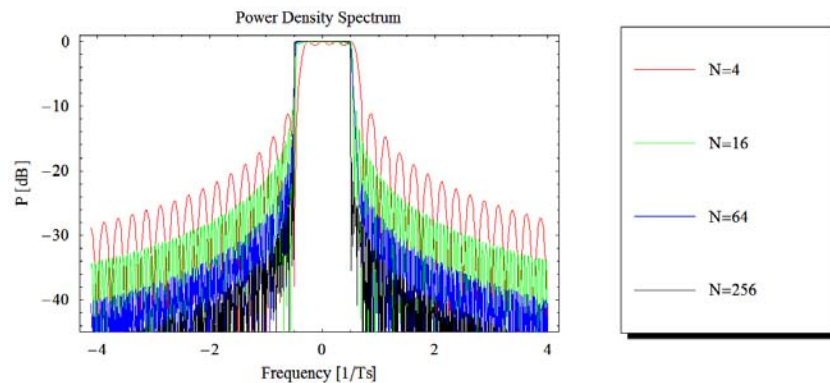
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OFDM power density spectrum

- Out-of band emissions due to rectangular symbol window
- decrease as N increases



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OFDM symbol

- subcarriers modulated with a complex symbol
 - changes the phase (frequency) during transmission
 - changes amplitude, if MQAM with $M > 4$
 - narrowband transmission
 - the information is packed in the frequencies used
 - frequencies remain constant in $[0, T]$
- the sum of $N' \leq N$ modulated subcarriere signals in time $T = NT_s$ is an OFDM symbol
 - Orthogonal Frequency Domain Multiplexing

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- multipath components of an OFDM symbol arrive with different timing, but same frequencies
- assume delay spread τ_s , transmission of a single OFDM symbol,
 - received in time $[0, T + \tau_s]$
 - multipath components do not change the frequency content of the signal
 - reception can be based on any period of length T within $[0, T + \tau_s]$
- when transmitting multiple OFDM symbols, Cyclic Prefix (CP) of length $\geq \tau_s$ added in front of OFDM symbol
 - at receiver, CP dropped \Rightarrow all Inter-Symbol Interference vanishes
- movement (Doppler shift) changes the frequency content of the signal \Rightarrow Inter-Carrier Interference

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DFT and FFT

- sampling the subcarrier waveforms at intervals of T_s one gets the sample representation of the subcarriers, the Hermitian conjugate of a DFT matrix M , e.g. $N = 4$ sampled at $t = -2, -1, 0, 1$:

$$\mathbf{M} = \frac{1}{2} \begin{bmatrix} 1 & j & -1 & -j \\ 1 & 1 & 1 & 1 \\ 1 & -j & 1 & j \\ 1 & -1 & 1 & -1 \end{bmatrix}$$

- by convention, OFDM transmission is defined as *inverse* Fourier transform \Rightarrow Hermitian conjugate
- the sampled subcarrier waveforms are columns in \mathbf{M}^H , conjugated rows in \mathbf{M} .

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- often $N \times N$ DFT matrix expressed with permuted columns (subcarriers) in the form with matrix elements $m_{ik} = e^{-2\pi j (i-1)(k-1)/N}$, e.g.:

$$\mathbf{M} = \frac{1}{2} \begin{bmatrix} 1 & 1 & 1 & 1 \\ 1 & j & -1 & -j \\ 1 & -1 & 1 & -1 \\ 1 & -j & 1 & j \end{bmatrix}$$

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- time domain samples of an OFDM symbol are Inverse Discrete Fourier Transforms of subcarrier symbols:

$$\mathbf{s} = \mathbf{M}^H \begin{bmatrix} x_1 \\ x_2 \\ \vdots \\ x_N \end{bmatrix}$$

- some of the highest & lowest frequency subcarriers may be left unused
- with $N = 2^m$, the DFT can be performed with Fast Fourier Transform (FFT)
 - $N \log_2 N$ multiplications required to construct the DFT
- at receiver, DFT or FFT is used to recover the symbols transmitted on the subcarriers

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OFDM summary

- using a discrete (or fast) Fourier transform, modulation is performed in frequency domain
- using a cyclic prefix removes all inter-symbol interference
- added overhead from transmitting CP
- perfect removal of multipath-generated interference
- in ADSL, OFDM (a.k.a. discrete multitone) is used
- OFDM is the modulation used in most recent mobile standards
 - WiMAX
 - 3G LTE